

CLAIMS:

1. A signal comprising a runlength limited (RLL) encoded binary d,k channel bitstream, wherein parameter d defines a minimum number and parameter k defines a maximum number of zeroes between any two ones of said bitstream or vice versa, comprising
 - 5 a number of sections of respectively N successive RLL channel bits, called RLL rows, each RLL row representing a parity-check code-word, called row parity-check code-word, in which a so-called row-based parity-check constraint for said RLL row has been realized; characterized in that
 - K sections of respectively N successive channel bits, called column parity-check rows, are
 - 10 located at predetermined positions of a group of M RLL rows, K , N and M being integer values, said column parity-check rows comprising a plurality of column parity-check enabling channel words,
 - wherein each of said column parity-check enabling channel words realizes a so-called column-based parity-check constraint for all so-called corresponding segments of at least said
 - 15 M RLL rows of said group that correspond to a specific column parity-check enabling channel word, hereby constituting a column parity-check codeword.
2. A signal according to claim 1, characterized in that the number K of column parity-check rows is at least 2.
- 20 3. A signal according to claim 1, characterized in that each one of said row parity-check code-words comprises a row parity-check enabling channel word being appended to RLL encoded user data without itself containing user data.
- 25 4. A signal according to claim 1, characterized in that each one of said row parity-check code-words comprises a row parity-check enabling channel word being encoded user data.

5. A signal according to claim 1, characterized in that each of said row parity-check code-words comprises a parity-check bit p_{2H} realizing the row-based parity-check constraint

$$p_{2H} = \text{Mod} \left[\sum_{i=0}^{N-1} i \cdot b_i, 2 \right]$$

5 which is at an encoder set to a predetermined value, known at a decoder, that is, it is either set to 0 or to 1, with b_i being successive bits of said d,k channel bitstream of an RLL row.

6. A signal according to claim 1, characterized in that said corresponding segments of RLL rows being respectively a certain bit at a predetermined position of each of
10 said RLL rows and a respective single-bit wide column parity-check enabling channel word or words is or are located at the same position of each of said K column parity-check rows.

7. A signal according to claim 6, characterized in that the number K of column parity-check rows is an integer value with $K \geq 1/R$, with R being the code rate of the RLL
15 code with RLL constraints d and k .

8. A signal according to claim 7, characterized in that each of said single-bit wide column parity-check enabling channel words is an encoded symbol of a parity-check bit p_{2V} realizing the column-based parity-check constraint

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$$p_{2V} = \text{mod} \left[\sum_{l=0}^{M+K-1} b_l, 2 \right]$$

which is at the encoder set to a predetermined value, known at the decoder, that is, either set to 0 or to 1, with b_l being bits of the RLL rows of said group at a certain position, wherein the bits of said symbol are spread over said K column parity-check rows, one bit per row.

25 9. A signal according to claim 8, characterized in that said symbol is selected from a number of different symbols realizing said column-based parity-check constraint p_{2V} in order to realize the d,k constraints of the RLL code within said column parity-check rows as well.

30 10. A signal according to claim 6, characterized in that said single-bit wide column parity-check enabling channel words are located at each channel bit position of a

column parity-check row, hereby constituting at every bit-wide column a column parity-check codeword.

11. A signal according to claim 6, characterized in that said single-bit wide
5 column parity-check enabling channel words are located at every second channel bit position only, hereby constituting at every second bit-wide column a column parity-check codeword.

12. A signal according to claim 11, characterized in that channel bits between said
10 second bit positions are used as merging bits in order to realize said d, k constraints of said RLL code and/or any desired spectral property of the code like DC-control.

13. A signal according to claim 1, characterized in that said predetermined
position of a row parity-check enabling channel word is at the end of an RLL row.

15 14. A signal according to claim 1, characterized in that said K column parity-check rows are arranged successively.

15. A signal according to claim 1, characterized in that said M RLL rows are
arranged successively.

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16. A signal according to claim 1, characterized in that said predetermined
position of said K column parity-check rows is at the end of said group.

17. A signal according to claim 1, characterized in that said predetermined
25 position of said K column parity-check rows is in front of said group.

18. A signal according to claim 1, characterized in that said predetermined
position of said K column parity-check rows is within said group.

30 19. A signal according to claim 1, characterized in that the number K of column
parity-check rows is two,
each of said column parity-check rows is divided into segments of at least two
types, and in the case of two types, of more than one successive channel bits of alternating

segment width N_1 or N_2 , N_1 and N_2 being integer values, N_1 being the width of the first column parity-check segment and N_2 being the width of the second parity-check segment,

wherein in each column parity-check row only every second segment is a column parity-check enabling channel word and

5 wherein only one of both column parity-check rows starts with a column parity-check enabling channel word,

whereas in the other column parity-check row the first column parity-check enabling channel word is at the second segment position.

10 20. A signal according to claim 19, characterized in that within each column parity-check row the so-called merging segments in front of or behind a column parity-check enabling channel word do not contain any user data, but are designed such to realize said d, k constraints of said RLL code and/or any desired spectral property of the code like DC-control.

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21. A signal according to claim 19, characterized in that in the first one of both column parity-check rows the parity-check information of each column parity-check enabling channel word realizes said parity-check constraint only for said column parity-check enabling channel word in addition to said corresponding segments of said M RLL rows of said group.

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22. A signal according to claim 21, characterized in that in the second one of both column parity-check rows the parity-check information of each column parity-check enabling channel word realizes said parity-check constraint only for said column parity-check enabling channel word in addition to said corresponding segments of said M RLL rows of said group.

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23. A signal according to claim 21, characterized in that in the second one of both column parity-check rows the parity-check information of each column parity-check enabling channel word realizes said parity-check constraint for said column parity-check enabling channel word as well as the corresponding merging segment of said first column parity-check row in addition to said corresponding segment of said M RLL rows of said group.

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24. A signal according to claim 19, characterized in that said segment or channel word widths N_1 and N_2 are in the range of $d \leq N_{1,2} \leq k$.

25. A signal according to claim 21 or 22, characterized in that said parity-check constraint is

$$V = \text{Mod} \left[\sum_{j=1}^{M+1} w_j, q_1 \right]$$

being set to a predetermined value at the encoder, known at the decoder, and is preferably set

5 to 0, wherein j is a unique index associated with each RLL row for $1 \leq j \leq M$ and an index associated with the actual column parity-check row for $j = M + 1$, and

wherein w_j is a unique index associated with each word W_j which defines one of a number of possible d, k constrained sequences of said segment width (N_1, N_2) , wherein such a word W_j is comprised in each corresponding segment.

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26. A signal according to claim 23, characterized in that said column based parity-check constraint is

$$V = \text{Mod} \left[\sum_{j=1}^{M+2} w_j, q_1 \right]$$

being set to a predetermined value at the encoder, known at the decoder, and is preferably set

15 to 0, wherein j is a unique index associated with each RLL row for $1 \leq j \leq M$ and with each column parity-check row for $j = M + 1, M + 2$, and wherein w_j is a unique index associated with each word W_j which defines one of the a number of possible d, k constrained sequences of said segment width (N_1, N_2) , wherein such a word W_j is comprised in each corresponding segment.

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27. A signal according to claim 25 or 26, characterized in that said unique index w_j is

$$w_j = \sum_{i=0}^{N_{1,2}-1} b_i^j \cdot N_d(i)$$

wherein b_i^j denotes bit number i of word W_j in row j and wherein $N_d(i)$ is the number of

25 possible d, k constrained sequences of length i .

28. A signal according to claim 1, characterized by a waveform comprising said d, k channel bitstream, wherein said waveform transitions between two states (land, pit) whenever a one occurs in said d, k channel bitstream and keeps its actual state whenever a

30 zero occurs in said d, k channel bitstream or vice versa.

29. A storage medium storing a signal according to any one of claims 1 to 28.

30. A storage medium according to claim 29, characterized in that said storage
5 medium is a recorded optical, magnetic, or magneto-optical disc or recoded magnetic tape.

31. A method for encoding a stream of user data bits comprising the steps of:
runlength limited (RLL) encoding said stream of user data bits into a binary d,k channel
bitstream comprising a number of sections of respectively N successive RLL channel bits,
10 called RLL rows, wherein parameter d defines a minimum number and parameter k defines a
maximum number of zeroes between any two ones of said bitstream or vice versa,
each RLL row representing a parity-check code-word, called row parity-check code-word in
which a so-called row-based parity-check constraint for said RLL row has been realized,
characterized by the further step of
15 generating K sections of respectively N successive channel bits, called column parity-check
rows, at predetermined positions of a group of M RLL rows, K , N and M being integer
values, said column parity-check rows comprising a plurality of column parity-check
enabling channel words,
wherein each of said column parity-check enabling channel words realizes a so-called
20 column-based parity-check constraint for so-called corresponding segments of at least said M
RLL rows of said group that correspond to a specific column parity-check enabling channel
word, hereby constituting a column parity-check codeword.

32. A method according to claim 31, characterized by generating a signal
25 according to any one of claims 1 to 30.

33. A device for encoding a stream of user data bits comprising:
encoding means for runlength limited (RLL) encoding a stream of user data bits into a binary
 d,k channel bitstream comprising a number of sections of respectively N successive RLL
30 channel bits, called RLL rows, wherein parameter d defines a minimum number and
parameter k defines a maximum number of zeroes between any two ones of said bitstream or
vice versa,

wherein each RLL row represents a parity-check code-word, called row parity-check code-word in which a so-called row-based parity-check constraint for said RLL row has been realized,

characterized in that

- 5 said encoding means being designed for generating K sections of respectively N successive channel bits, called column parity-check rows, at predetermined positions of a group of M RLL rows, K , N and M being integer values, said column parity-check rows comprising a plurality of column parity-check enabling channel words, wherein each of said column parity-check enabling channel words realizes a so-called column-based parity-check
- 10 constraint for all so-called corresponding segments of at least said M RLL rows of said group that correspond to a specific column parity-check enabling channel word, hereby constituting a column parity-check codeword.

34. A device according to claim 33, characterized in that said device comprising
- 15 means for performing a method according to claims 31 or 32 in order to generate a signal according to any one of claims 1 to 30.

35. A method for decoding a signal according to any one of claims 1 to 30 or a signal being encoded according to a method of claim 31 or 32, comprising the steps of:
- 20 checking for each RLL row a so-called row-based parity-check constraint,
checking for each column parity-check segment of said column parity-check rows a so-called column-based parity-check constraint along all corresponding segments of at least said M RLL rows that correspond to said column parity-check enabling channel word,
and
- 25 determining an erroneous channel word based on said checking steps.

36. A method according to claim 35, wherein said determining step includes locating an erroneous segment at a crossing point of
- a) an erroneous RLL row that violates said row-based parity-check constraint for
- 30 said RLL row and
- b) an erroneous column comprising all corresponding segments that correspond to a specific column parity-check enabling channel word, wherein said column violates said column-based parity-check constraint.

37. A method according to claim 36, wherein an located erroneous segment is corrected if a single erroneous segment occurs.

38. A method according to claim 35, wherein said determining step is further
5 based on channel side-information if more than a single erroneous segment occurs.

39. A method according to claim 38, wherein said channel side-information is phase-error information of bit transitions in the channel words of the segments at said crossing points.

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40. A method according to claim 39, wherein a phase-error with the largest absolute value is determined and the corresponding one-bit of the d,k channel bitstream is shifted by one bit position.

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41. A method according to claim 35, wherein said signal comprises said column parity-check enabling channel words at every second channel bit position only, and wherein said determining step includes, upon detecting a first erroneous column, the step of deciding whether another erroneous column is positioned to the left or to the right of said first erroneous column.

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42. A method according to claim 41, wherein said decision step is based on channel-side information.

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43. A method according to claim 35, wherein said signal comprises segments of more than one successive channel bits of alternating segment width N_1 or N_2 , and wherein a single-bit transition-shift error is determined internal of such a segment, namely

a) a transition-shift error is determined from bit position i to the right to bit position $i+1$, if the detected column-based parity-check constraint is detected as

$$V_{as-detected} = N_d(i+1) - N_d(i)$$

30

and

b) a transition-shift error is determined from bit position i to the left to bit position $i-1$, if the detected column-based parity-check constraint is detected as

$$V_{as-detected} = N_d(i-1) - N_d(i)$$

wherein $N_d(i-1)$, $N_d(i)$, $N_d(i+1)$ are the numbers of possible d, k constrained sequences of length $i-1$, i , $i+1$, respectively.

44. A method according to claim 35, wherein said signal comprises segments of more than one successive channel bits of alternating segment width N_1 or N_2 , and wherein a single-bit transition-shift error is determined crossing the left boundary of such a segment, namely

a) a transition-shift error is determined from the last bit position of the previous segment to the first bit position of the present segment, if the detected column-based parity-

10 check constraint is detected for the present column as

$$V_{\text{as-detected, present}} = +N_d(0)$$

and if the detected column-based parity-check constraint is detected for the previous column as

$$V_{\text{as-detected, previous}} = -N_d(N_{1,2} - 1)$$

15 or

b) a transition-shift error is determined from the first bit position of the present segment to the last bit position of the previous segment, if the detected column-based parity-check constraint is detected for the present column as

$$V_{\text{as-detected, present}} = -N_d(0)$$

20 and if the detected column-based parity-check constraint is detected for the previous column as

$$V_{\text{as-detected, previous}} = +N_d(N_{1,2} - 1)$$

wherein $N_d(0) = 1$ and $N_d(N_{1,2} - 1)$ is the number of possible d, k constrained sequences of length $N_{1,2} - 1$.

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45. A method according to claim 35, wherein said signal comprises segments of more than one successive channel bits of alternating segment width N_1 or N_2 , and wherein a single-bit transition-shift error is determined crossing the right boundary of such a segment, namely

30 a) a transition-shift error is determined from the last bit position of the present segment to the first bit position of the subsequent segment, if the detected column-based parity-check constraint is detected for the present column as

$$V_{\text{as-detected, present}} = -N_d(N_{1,2} - 1)$$

and if the detected column-based parity-check constraint is detected for the subsequent column as

$$V_{\text{as-detected, subsequent}} = +N_d(0)$$

or

- 5 b) a transition-shift error is determined from the first bit position of the subsequent segment to the last bit position of the present segment, if the detected column-based parity-check constraint is detected for the present column as

$$V_{\text{as-detected, present}} = +N_d(N_{1,2} - 1)$$

- 10 and if the detected column-based parity-check constraint is detected for the subsequent column as

$$V_{\text{as-detected, subsequent}} = -N_d(0)$$

wherein $N_d(0) = 1$ and $N_d(N_{1,2} - 1)$ is the number of possible d, k constrained sequences of length $N_{1,2} - 1$.

- 15 46. A method according to claim 43, wherein a segment with a determined single-bit transition-shift error is corrected by being replaced by a segment having said unique index

$$w_j = w'_j - V_{\text{as-detected}}$$

wherein w'_j is an as-detected index of said segment to be replaced, wherein

$$w'_j = \sum_{i=0}^{N_{1,2}-1} b'_i \cdot N_d(i)$$

- 20 wherein b'_i denotes as-detected bit-value for the bit with number i of said segment in row j and wherein $N_d(i)$ is the number of possible d, k constrained sequences of length i .

47. A device for decoding a signal according to any one of claims 1 to 30 or a signal being encoded according to a method of claim 31 or 32, comprising:

- 25 - parity-check means for checking for each RLL row a row-based parity-check constraint, and for checking for each column parity-check enabling channel word of said column parity-check rows a so-called column-based parity-check constraint along all corresponding segments of at least said M RLL rows that correspond to said column parity-check enabling channel word, and
- 30 - determining means for determining an erroneous channel word based on the result of said parity-checking.

48. A device according to claim 47, wherein said device comprising means for performing a method according to any one of claims 35 to 46.

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